Supplement to The Wireless Electrical Trader, January 2, 1943

"TRADER" SERVICE SHEET

AC/DC TRANSPORTABLE



OMPACT in design, the Pye "Mite is a 3-valve (plus rectifier) 2-band TRF receiver for operation on AC or DC mains of 200-250 V, 25-100 C/S in the case of AC.

Its style follows midget receiver practice, and a line cord mains lead is used, but the circuit is more elaborate, and the components used are British.

Release date: 1939.

CIRCUIT DESCRIPTION

Input from attached aerial via isolating condenser C1 and coupling coil L1 to single tuned circuit L2, L3, C16, which precedes variable-mu RF amplifying valve (V1, Mullard Amerty 6K7G). Gain control by dual-action potentiometer which works in conjunction with R4 to form a potential divider across the HT circuit. As the control is advanced, the slider moves downwards in our circuit

diagram, reducing the damping effect acros the aerial circuit, at the same time reducing the positive potential of V1 cathode, since the slider is connected to chassis.

R3 limits the excursion of the cathode potential in the negative direction when the slider reaches the bottom of R1 (maximum gain), so that the cathode can never reach chassis potential. Reaction adds a third phase of action, and occurs as a result of the inherent instability of the receiver when the gain control is at maximum.

Tuned-anode coupling by L4, L5, C18, via C6, between V1 and another RF pentode valve (V2, Mullard Amerty 6J7G) which operates as detector on the grid leak system with C6 and R6. RF filtering by C8 in anode circuit.

Resistance-capacity coupling by R8, C9 and R10, R11, via grid stopper R9, between V2 and pentode output valve (V3, Mullard Amerty 25A6G). Fixed tone correction by C11 in anode circuit.

When the receiver is operating on AC mains, HT current is supplied by IHC half-wave rectifying valve V5, which on DC mains behaves as a low resistance. The valve may be a 25Z4G, as shown by the solid lines in our diagram, or 25Z6G, with its two sections connected in parallel as shown by the additional dotted portion in our diagram. The holder is so wired that either valve may be used.

Valve heaters, together with line cord ballast resistance R15, are connected in series across mains input. An additional series resistance R16 may be inserted in place of the high voltage link for use on

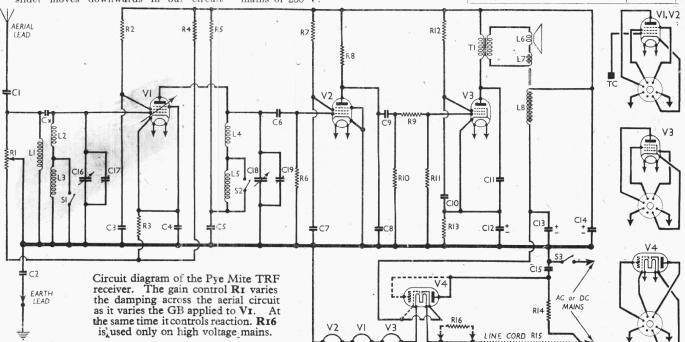
mains of 250 V

COMPONENTS AND VALUES

	CONDENSERS	Values (μF)
C1 C2	Mains isolating condensers {	0·001 0·05
$\overline{\text{C3}}$	VI SG decoupling	0.1
C4	VI Cathode by-pass	0.1
C5	VI anode decoupling	0.1
C6	V2 CG condenser	0.00015
C7	V2 SG decoupling	0.1
C8	RF by-pass	0.0003
C9	V2 to V3 AF coupling	0.01
C10	V3 SG decoupling	0.1
C11	Fixed tone correcter	0.01
C12*.	V3 cathode by-pass	10.0
C13* C14*	HT smoothing condensers {	16·0 8·0
C15	Mains RF by-pass	0.01
C16†	Aerial circuit tuning	
C17‡	Aerial circ. MW trimmer	
C18†	V1 anode circuit tuning	
C19‡	V1 anode circ. MW trim- mer	
Cx	Aerial "top" coupling	Very low

* Electrolytic. † Variable. ‡ Pre-set.

RESISTANCES.	Values (ohms)	
R1		10,000 100,000 450 50,000 15,000 1,000,000 250,000 40,000 1,000,000 1,000,000 1,000,000 10,000 530 60



OTHER COMPONENTS	Approx. Values (ohms)
$ \begin{array}{c c} L1 \\ L2 \\ L3 \\ L4 \\ L6 \\ L6 \\ L7 \\ L8 \\ \end{array} $	11·0 3·0 11·0 3·5 11·0 2·0 0·1 900·0 200·0 0·1

VALVE ANALYSIS

Valve voltages and currents given in the table below are those given in the makers' manual, and they represent conditions to be expected in an average receiver when it is operating on AC mains of 220 V, with the gain control at maximum, but with no signal input.

Voltages were measured with 1,000 ohms-per-volt meter, chassis being the negative connection.

Valve	Anode	Anode	Screen	Screen
	Voltage	Current	Voltage	Current
	(V)	(mA)	(V)	(mA)
V1 6K7G	130	4·5	73	1.2
V2 6J7G	40	0·7	25	
V3 25A6G V4 25Z4G	203 240†	34·0 53·0*	139	6.4

† Cathode to Chassis, DC. * Cathode current, DC.

DISMANTLING THE SET

Removing Chassis.—Remove the three control knobs (recessed grub screws); remove the two round-head screws holding the chassis to the bottom of the cabinet. Chassis can now be withdrawn to the extent of the speaker leads, which is sufficient for normal purposes.

To free chassis entirely, unsolder the leads from the panel on the speaker input transformer.

former.

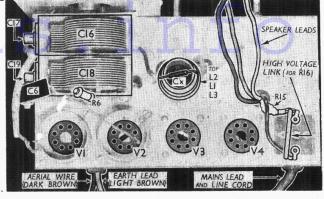
former.

When replacing, connect the speaker leads as follows, numbering the tags from top to bottom: 1 and 2 (joined together), black; 3, no external connection; 4, yellow; 5, red.

Removing Speaker.—Remove chassis, then remove the two nuts (with lock-washers) holding the speaker to the sub-baffle.

When replacing, the speaker should be on the right. The leads, if unsoldered, should be connected as described above.

Plan view of the chassis. Cx is made up enamelled wire. The wire link which takes the place of R16 in low voltage models is clearly indicated. and R6 are mounted on a panel fixed to one endplate of the gang assembly.



GENERAL NOTES

Switches .- S1, S2 are the waveband switches, in a single rotary unit beneath the chassis. The unit is indicated in our under-chassis view and shown in detail in the diagram below. Both switches are closed on MW, and open on LW.

83 is the QMB mains switch, ganged with the gain control R1.



Diagram the wave band switch unit, as seen from the rear of the underside of the chassis.

Coils.-L1-L3 are wound on an unscreened tubular former mounted on the chassis deck. A small condenser, Cx, composed of enamelled wire, fits across the top of the unit. L4, L5 are on a similar former mounted horizontally beneath the chassis.

External Speaker.-No provision is made for this, but a high-impedance

(about 5,000 O) type could be connected, via a pair of high-voltage rating condensers of about 0.1 μF or more each, across the primary of transformer T1.

Heater Circuit Ballast .- The main heater circuit ballast resistance R15 is of the "line cord" type. It is housed in the mains lead, and is satisfactory for mains of 200-235 V. A subsidiary resistance of 60 O, rated at 6 watts, should be connected in place of the high-voltage link shown in our plan view for operation on mains of 240-250 V.

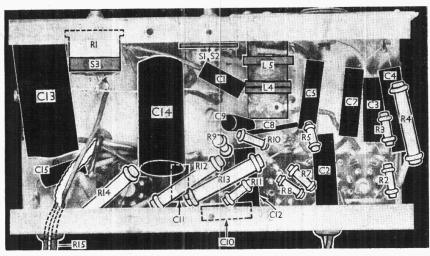
Valve V4.—The makers' information shows this as a 25Z6G, which is of the voltage-doubler type, with its two halves connected in parallel. Our sample was fitted with the half-wave type 25Z4G. Either valve may be used without rewiring the holder, the only difference being that pins 3 and 4, although wired, are not used. The valve base diagram given overleaf for V4 is correct for the 25Z4G if the left-hand section of the valve (shaded in our diagram) is omitted.

Chassis Divergencies.—The second V3 CG resistance R11 was not shown in the makers' circuit diagram, but was present in our chassis. C10 was shown in the makers' diagram returned to chassis, instead of to V3 cathode. Also, C8 was given as a 0.003 μF tubular paper condenser, whereas in our chassis it was a 0.0003 μ F silvered mica type. A different speaker unit from the original appears to have been used in our chassis, as the makers give DC resistance values as follow: **T1**, pri. 128 O; **L6**, 5 O; **L8**, 629 O. The serial number of our chassis was MCF 1393.

CIRCUIT ALIGNMENT

With the gang at minimum, the pointer should be horizontal. Adjustment is a simple matter, as the pointer is soldered to a disc having a peg which is a friction fit in a hole in the end of the gang spindle. Connect signal generator leads to the ends of the A and E leads.

Switch set to MW (anti-clockwise), tume to 200 m on scale, feed in a 200 m (1,500 KC/S) signal, and adjust C17 and C19 for maximum output, while adjusting the gain control for optimum results. Repeat the operation until no improvement can be obtained. There are no LW adjustments, but the calibration should be checked at several points on both wave-



The line cord resistance R15 is indicated as it runs from the hole in the chassis deck and into the mains lead. A diagram showing the waveband switch unit in detail appears above in column 2.